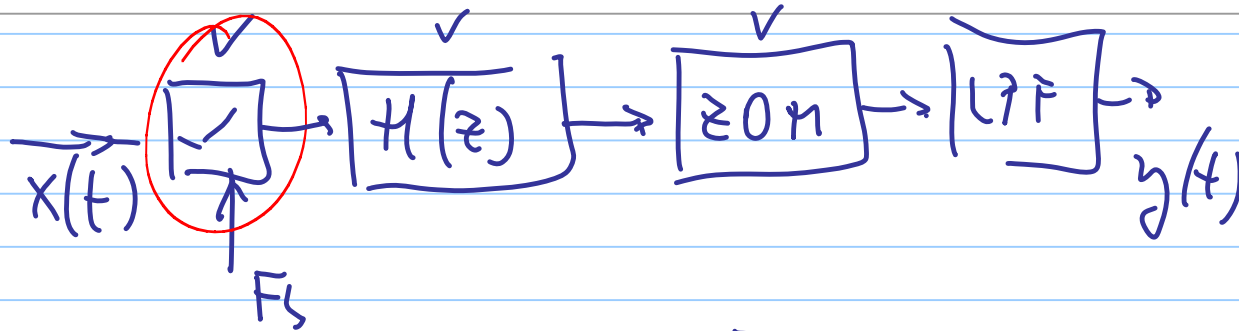


PROB. 2.5 / 2.6



$$Y(F) = \left(\frac{1}{F_s} H(\omega) \Big|_{\omega = \frac{2\pi F}{F_s}} \frac{T_s e^{-j\frac{2\pi F T_s}{F_s}} \text{sinc}\left(\frac{F}{F_s}\right) \right) X(F)$$

(no aliasing)

$$Y(F) = \bar{H}(F) X(F)$$

Example:

$$\text{if } X(F) = A e^{j\alpha} \delta(F - F_0) + A e^{-j\alpha} \delta(F + F_0)$$

Since no aliasing

$$-\frac{F_s}{2} < F_0 < \frac{F_s}{2}$$

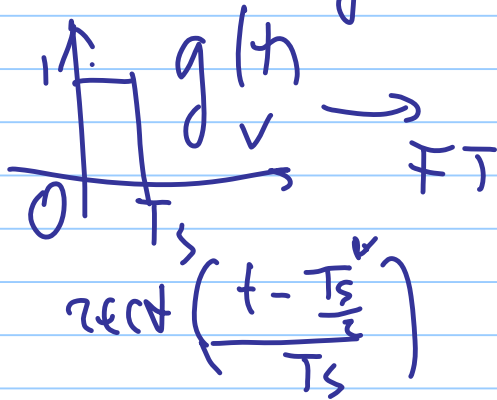
$$Y(F) = A e^{j\alpha} H(F_0) \delta(F - F_0) + \\ + A e^{-j\alpha} H(-F_0) \delta(F + F_0)$$



$$Y(F) = G(F) Y(w)$$

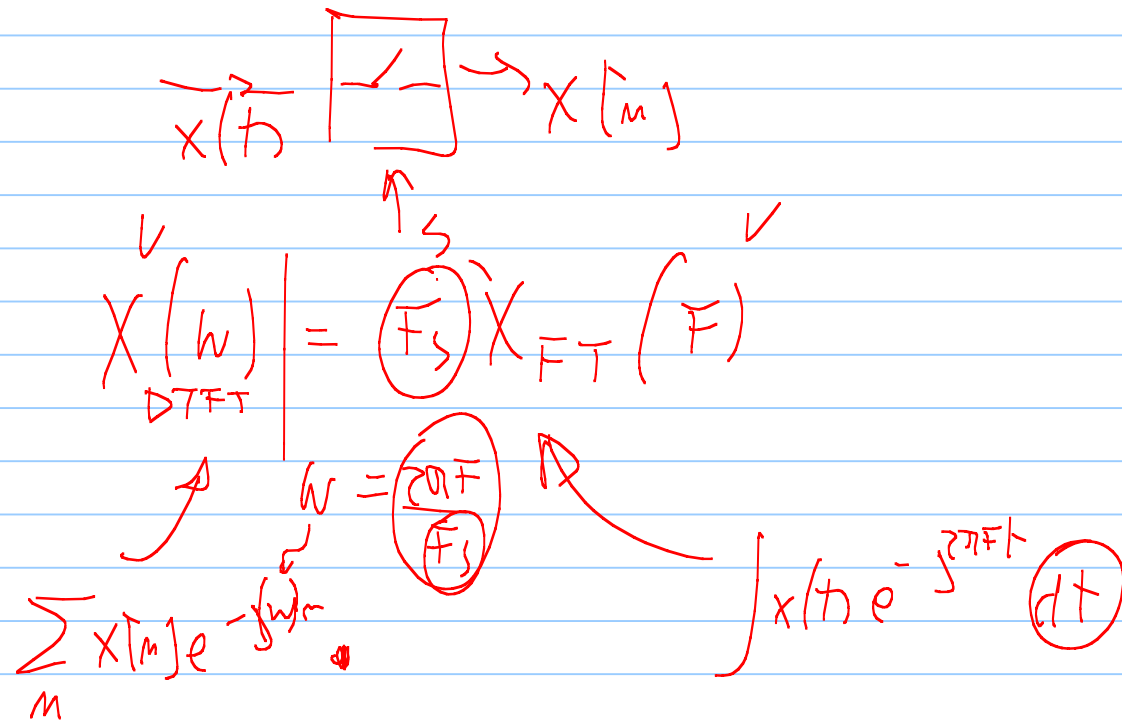
$$G(F) = \text{FT} \{ g(t) \}$$

$$w = 2\pi \frac{F}{f_s}$$



$$T_s \text{sinc}\left(\frac{F}{f_s}\right) e^{-j\pi F T_s}$$

if no aliasing



Example with aliasing

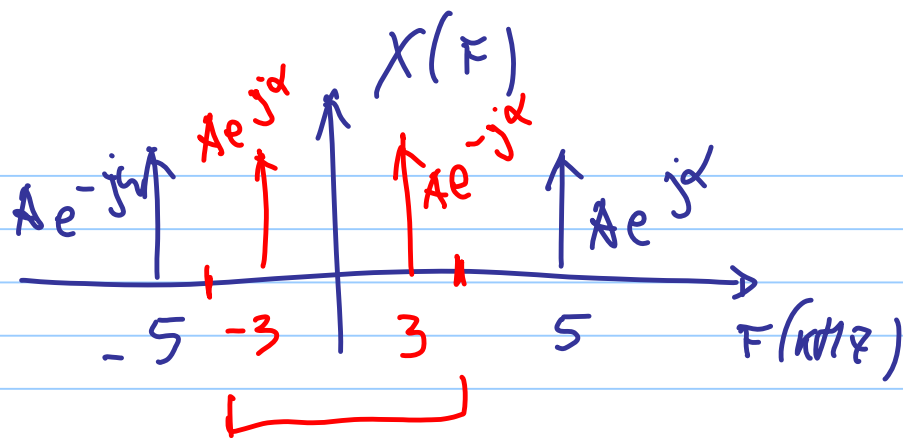
$$X(F) = A e^{j\alpha} \delta(F - F_0) + A e^{-j\alpha} \delta(F + F_0)$$

Ex:

$$F_0 = 5 \text{ kHz} \checkmark$$

$$F_s = 8 \text{ kHz} \checkmark$$

Step 1: find the aliased freq. F_1



$$X_1(F) = Ae^{-j\omega t} \delta(\underline{F-3000}) + Ae^{j\omega t} \delta(\underline{F+3000}) \quad \checkmark$$

the proceed as before

$$Y(F) = \bar{H}(3000) Ae^{-j\omega t} \delta(F-3000) - \dots$$

PROBLEM 3.2c

$$\text{DTFT}\{m x[n]\} = j \frac{d}{d\omega} X(\omega)$$

Since:

$$\textcircled{j} \frac{d}{d\omega} X(\omega) = j \frac{d}{d\omega} \sum_n x[n] e^{-j\omega n}$$

correct

$$= j \cancel{(\cancel{j})} \underbrace{\sum_n n x[n] e^{-j\omega n}}_{\text{DTFT}\{m x[n]\}}$$

Ex: DTFT of $\underbrace{0.8^n u[n]}_{x[n]} = ?$

$$\begin{aligned} X(\omega) &= \text{DTFT} \{ 0.8^n u[n] \} = \\ &= \frac{1}{1 - 0.8 e^{-j\omega}} \end{aligned}$$

apply the property .

Properties of DFT

$$\text{DFT} \{ x[(n-m)_N] \} = e^{-j\frac{2\pi}{N}km} X[k]$$

for all $k=0, \dots, N-1$

Ex:

$$X = [1, j, 0, -j] \quad N = 4$$

$$\text{DFT} \{ x[(n-2)_4] \} = ?$$

$$= e^{-j \frac{2\pi}{4} n k} X[k], \quad k = 0, \dots, 3$$

$$= (-1)^n X[k]$$

$$= [1, -j, 0, j] //$$

PROBLEM 3.3d

$$x[m] = \cos(0.25\pi m), m = 0, \dots, 7$$

$$= \frac{1}{2} e^{j\frac{\pi}{4}m} + \frac{1}{2} e^{-j\frac{\pi}{4}m}$$

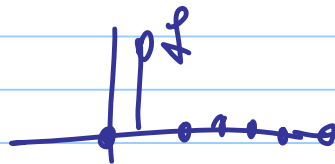
$$\begin{aligned} \text{DFT} \left\{ e^{j\frac{\pi}{4}m} \right\} &= \sum_{m=0}^7 e^{j\frac{\pi}{4}m} e^{-jk\frac{2\pi}{8}m} = \\ &= \sum_{m=0}^7 e^{j\frac{\pi}{4}(1-k)m} = (\text{geometric}) \end{aligned}$$

$$\sum_{n=0}^{\infty} \left(e^{j\frac{\pi}{4}(1-\kappa)} \right)^n =$$

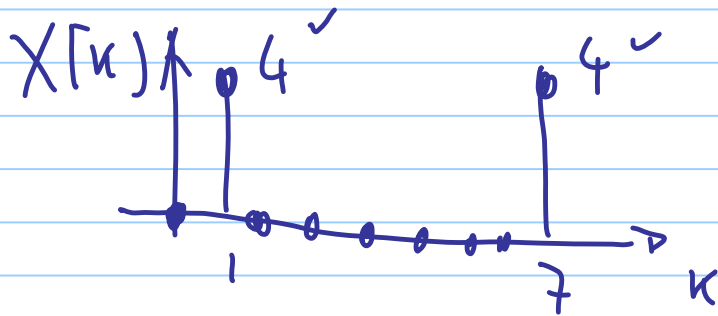
$$= \frac{1 - e^{j\frac{\pi}{4}(1-\kappa)} \delta^2}{1 - e^{j\frac{\pi}{4}(1-\kappa)}} = 0, \kappa \neq 1$$

$$= \delta \quad \text{if } \kappa = 1$$

$$= \delta \delta[\kappa - 1]$$



$$X[k] = \frac{8}{2} \delta[(k-1)_8] + \frac{8}{2} \delta[(k+1)_8]$$



PROBLEM 3.4 a, b

a) $x[(m-1)_4]$

b) $x[(m+3)_4]$

same

Since

$$x[(m-1)_4] = x[(m+4-1)_4]$$

$$= x[(m+3)_4]$$

$$x[(m+3)_4] = x[(m-4+3)_4] = x[(m-1)_4]$$

In fact, using DFT:

$$\text{DFT} \{ x[(m-1)_4] \} = e^{-j k \frac{2\pi}{4}} X[k]$$

$$\text{DFT} \{ x[(m+3)_4] \} = e^{j k \frac{2\pi}{4} 3} X[k]$$

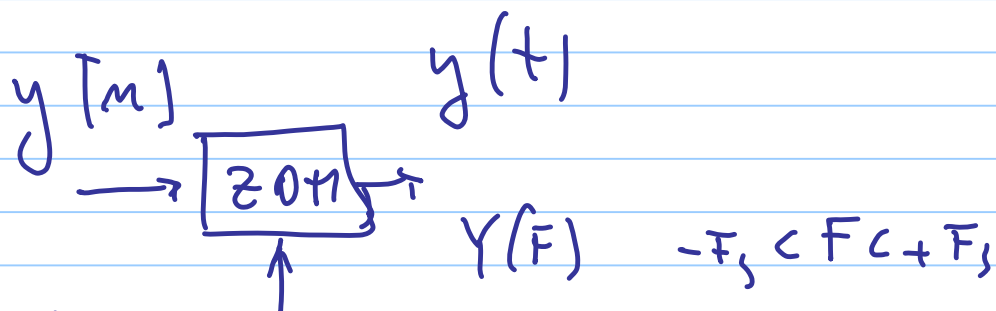
↑
same

Simp

$$(-j)^k = j^{3k}$$

$$j^{3k} = j^{2k} j^k = (-1)^k j^k = (-j)^k$$

PROBLEM 2.4



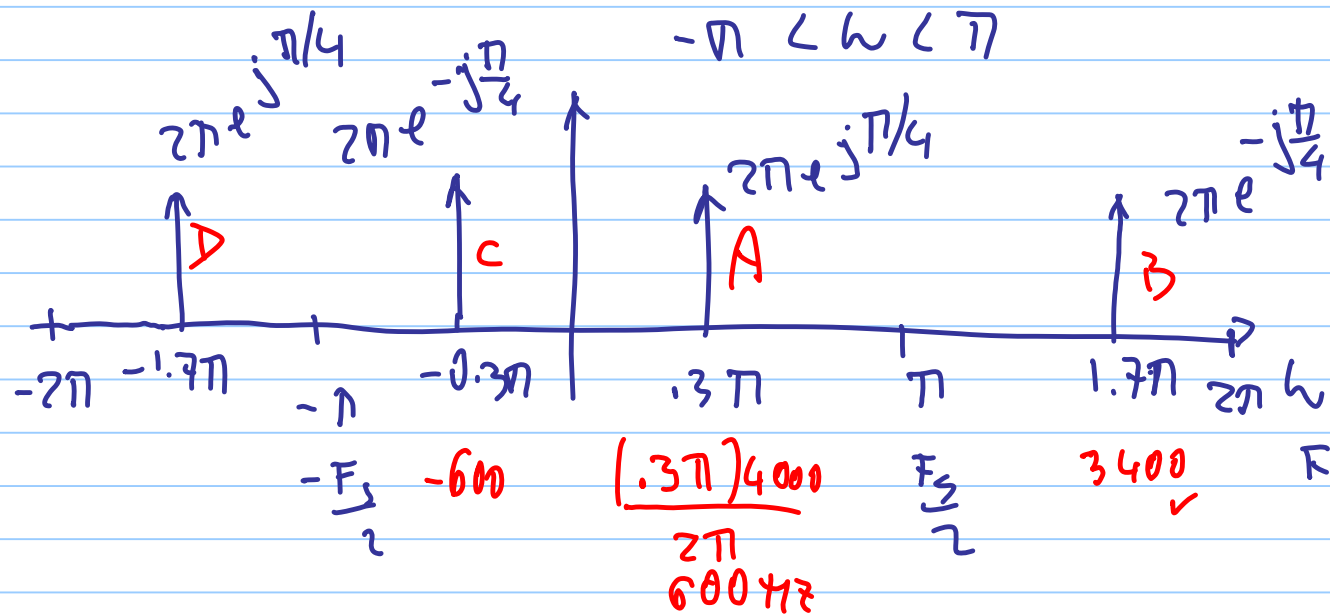
Recall: $F_s = 4\text{kHz}$

$$Y(F) = T_s e^{-j\pi \frac{F}{F_s}} \text{sinc}\left(\frac{F}{F_s}\right) Y(\omega) \Big|_{\omega = \dots}$$

$$= G(F) Y(\omega) \Big|_{\omega = 2\pi \frac{F}{F_s}}$$

$$Y(\omega) = \text{DTFT} \left\{ e^{j\frac{\pi}{4}} e^{j0.3\pi n} + e^{-j\frac{\pi}{4}} e^{-j0.3\pi n} \right\}$$

$$= 2\pi e^{j\frac{\pi}{4}} \delta(\omega - 0.3\pi) + 2\pi e^{-j\frac{\pi}{4}} \delta(\omega + 0.3\pi)$$



$$Y(F) = \frac{1}{4000} \underbrace{\left(e^{-j\pi \frac{600}{4000}} \right)}_{G(600)} \operatorname{sinc}\left(\frac{600}{4000}\right)$$

$$2\pi e^{j\frac{\pi}{4}} \delta\left(\frac{2\pi F}{4000} - 0.3\pi\right) + \dots$$

$$= G(600) 2\pi e^{j\frac{\pi}{4}} \frac{4000}{2\pi} \delta(F - 600) + \dots$$

Chapter 4

4.6 FIR Low pass

Passband $0 \rightarrow 10 \text{ kHz}$

STOP $> 11 \text{ kHz}$, $A = 50 \text{ dB}$

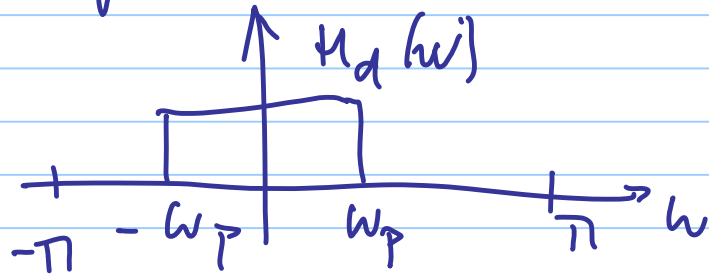
$$F_s = 44 \text{ kHz}$$

① Digital freq.

$$\omega_p = \frac{2\pi \cdot 10}{44} = \dots$$

$$\omega_{\text{stop}} = 2\pi \frac{11}{44} = \frac{\pi}{2}$$

② ideal filter $h_d[m]$



$$h_d[m] = \frac{1}{2\pi} \int_{-\omega_p}^{\omega_p} e^{j\omega m} d\omega$$

$$= \frac{\omega_p}{\pi} \operatorname{sinc}\left(\frac{\omega_p}{\pi} m\right) \quad -\infty < m < +\infty$$

③ Window: Blackman window

$$\Delta\omega \geq \frac{16\pi}{N}$$

$$\Delta\omega = \omega_{\text{stop}} - \omega_{\text{pass}} = \dots$$

$$N \geq \frac{16\pi}{\Delta\omega} \quad \text{choose } N \text{ odd}$$

$$N = 2L + 1$$

$$L = \text{true delay}$$

$$\textcircled{4} \quad h[m] = h_d[m-L] w_{\text{Blackman}}[m]$$

$$= \frac{w_p}{\pi} \operatorname{sinc}\left(\frac{w_p}{\pi}(m-L)\right) w_B[m]$$

$$m = 0, \dots, N-1$$

impulse response

$$y[m] = h[0]x[m] + h[1]x[m-1]$$

Typical Problem:

arbitrary

$$H(\omega) = \begin{cases} j|\omega| & |\omega| < \frac{\pi}{4} \\ 0 & |\omega| > \frac{1.2\pi}{4} \end{cases}$$

$A = 40 \text{ dB}$

Same kind of solution:

① ideal filter

$$h_d[m] = \frac{1}{2\pi} \int_{-\frac{\pi}{4}}^{\frac{\pi}{4}} j|\omega| e^{j\omega n} d\omega$$

Recall $|\omega| = \begin{cases} \omega & \text{if } \omega > 0 \\ -\omega & \text{if } \omega < 0 \end{cases}$

$$\textcircled{3} \quad h[n] = h_d[n-L] w_h[n]$$
$$n = 0, \dots, N-1$$

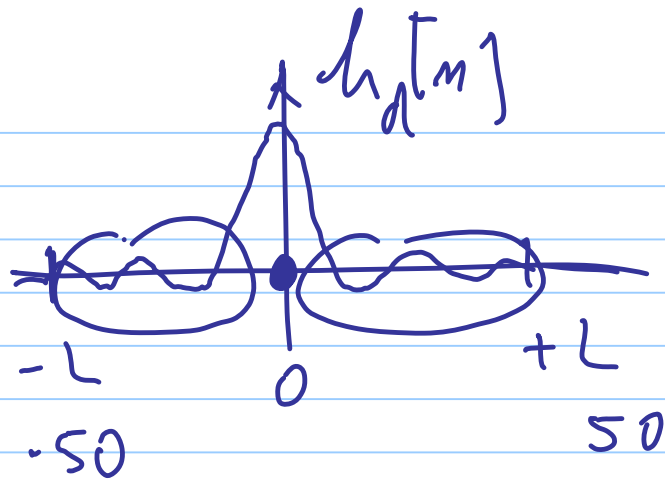
$$N \geq 100$$

choose $N = 101$ length

$L = \underline{\underline{50}}$ time delay

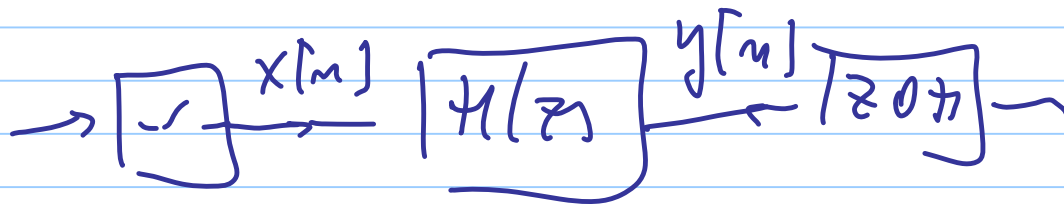
$$\Delta\omega \geq \frac{8\pi}{N}$$

$$\underline{\underline{N}} \geq \frac{8\pi}{\Delta\omega} = 100$$



$$N = 50 + 50 + 1 = 101$$

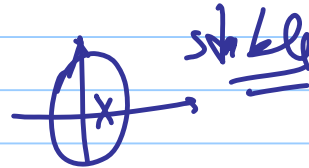
z-T ansform -

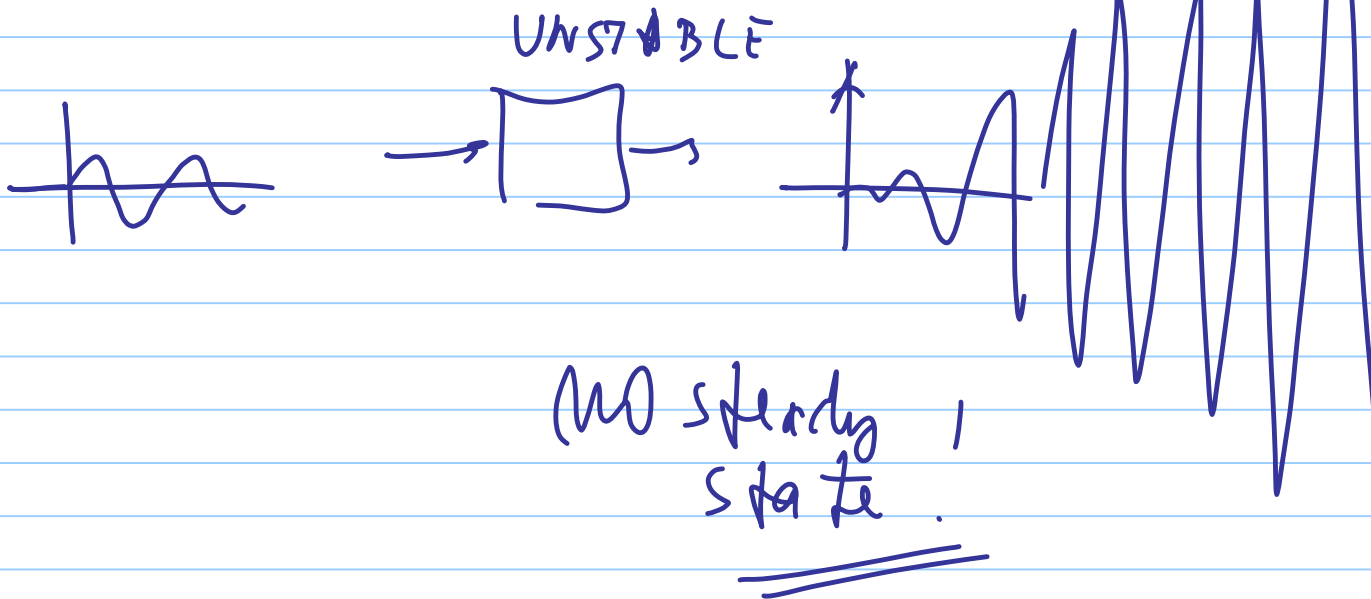
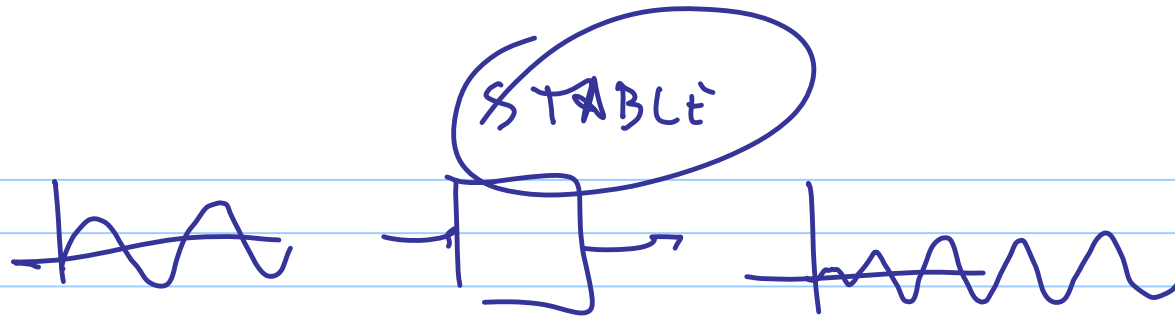


$$\text{let } y[n] = 0.6y[n-1] + x[n] + 2x[n-1] \quad \checkmark$$

$$H(z) = \frac{1 + 2z^{-1}}{1 - 0.6z^{-1}} = \frac{z + 2}{z - 0.6} \quad \checkmark$$

stable?

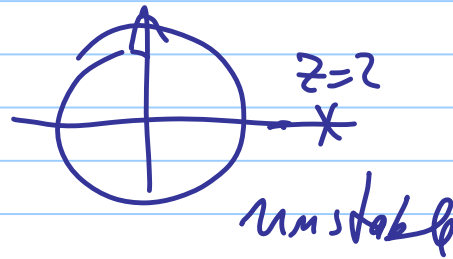




Ex: $y[n] = 2y[n-1] + x[n-1]$

$$H(z) = \frac{z^{-1}}{1 - 2z^{-1}} = \frac{1}{z - 2}$$

unstable



Freq. Response?

~~$H(\omega) = \frac{1}{e^{j\omega} - 2}$~~

it does not exist