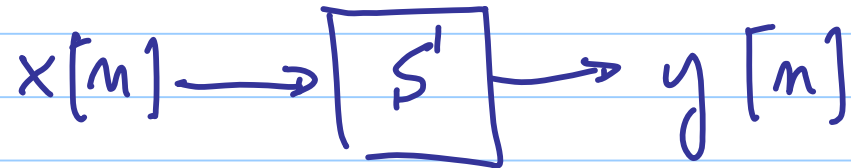


Recall: z-TRANSFORM AND FREQ. RESP.

Note Title

4/5/2011



Assume: S is Linear, Time Invariant
(LTI)

Typical: LINEAR DIFFERENCE EQ.

$$\begin{aligned} y[n] + a_1 y[n-1] + \dots + a_N y[n-N] &= \\ &= b_0 x[n] + b_1 x[n-1] + \dots + b_N x[n-N] \end{aligned}$$

Example: $N = \text{order}$, $N = 1$

$$\underline{y[m] + 0.8y[m-1] = 2x[m-1]}$$

$$a_1 = 0.8, \quad b_0 = 0, \quad b_1 = 2$$



Implementation by recursion:

EFFECT

$$y[m] = -0.8y[m-1] + 2x[m-1]$$

INPUT

CAUSAL

↑
CURRENT
OUTPUT

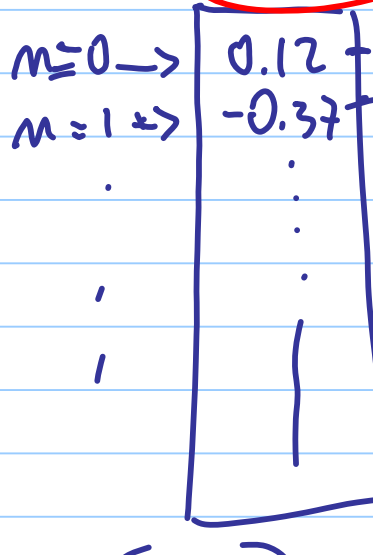
↑
PAST
OUTPUT

↑
PAST
INPUT

Observation: discrete time sequence

$x[n]$ = file with numbers

F_s



$n \rightarrow$ pointer in the file.

No time information

$x[n]$
↑
index

$X(\omega)$

no time inf.

$$\omega = \frac{2\pi f}{F_s} \quad \text{no dim.}$$

Observation on causality:

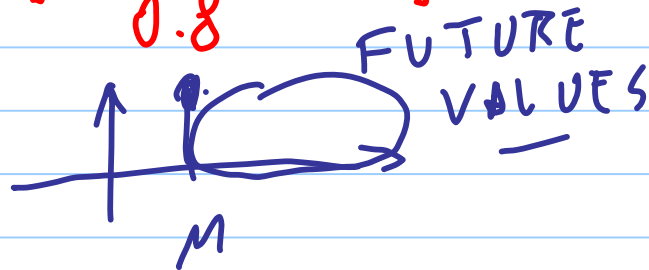
same LDE

$$y[n] + 0.8y[n-1] = 2x[n-1]$$

① $y[n] = -0.8y[n-1] + 2x[n-1]$ CAUSAL

② $y[n-1] = -\frac{1}{0.8}y[n] + \frac{2}{0.8}x[n-1]$

NOT CAUSAL



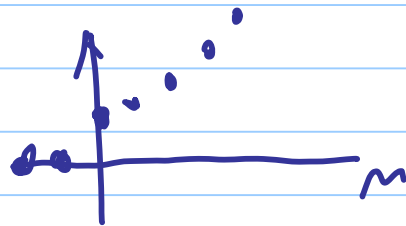
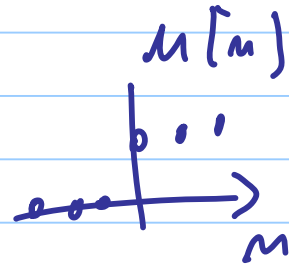
Definition: z-TRANSFORM

$$X(z) = \sum_{n=-\infty}^{+\infty} \{x[n]\} z^{-n}$$

$z = \text{complex variable}$

Ex: $x[n] = 2^n u[n]$

$$u[n] = \begin{cases} 1 & \text{if } n \geq 0 \\ 0 & \text{if } n < 0 \end{cases}$$



$$\begin{aligned}
 X(z) &= \sum_{n=-\infty}^{+\infty} 2^n u[n] z^{-n} = \sum_{n=0}^{+\infty} 2^n z^{-n} = \\
 &= \sum_{n=0}^{+\infty} (2z^{-1})^n = ? \quad (1)
 \end{aligned}$$

Recall: geometric series

$$\sum_{n=0}^{+\infty} \alpha^n = \begin{cases} \frac{1}{1-\alpha} & \text{if } |\alpha| < 1 \\ \text{not conv.} & \text{if } |\alpha| \geq 1 \\ \text{(NaN)} & \end{cases}$$

Apply geometric series to (1) : $|z|=2$

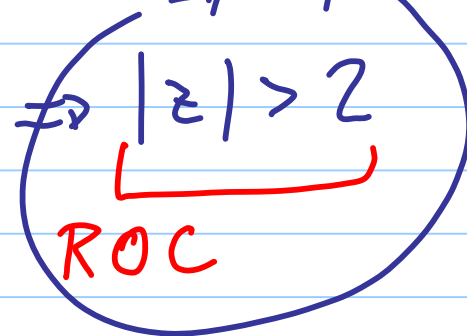
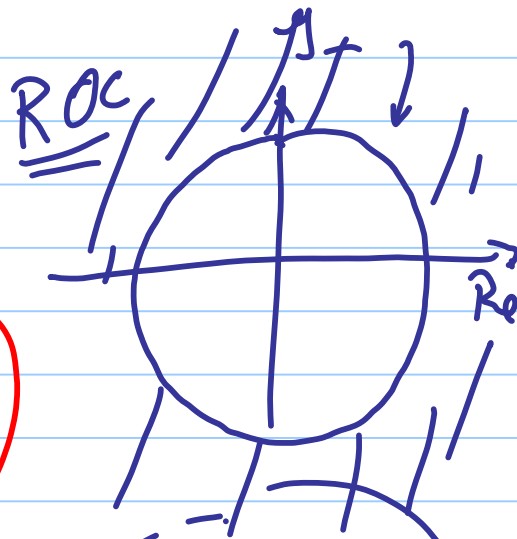
$$d = 2z^{-1},$$

the

$$X(z) = \frac{1}{1 - 2z^{-1}} = \frac{z}{z-2}$$

$$|d| < 1 \Rightarrow |2z^{-1}| < 1 \Rightarrow |z| > 2$$

ROC



Compare z-Transform and DTFT

Recall:

$$X(\omega) = \sum_{m=-\infty}^{+\infty} x[m] e^{-j\omega m}$$

↑
same as $X(z)$
 $z = e^{j\omega}$

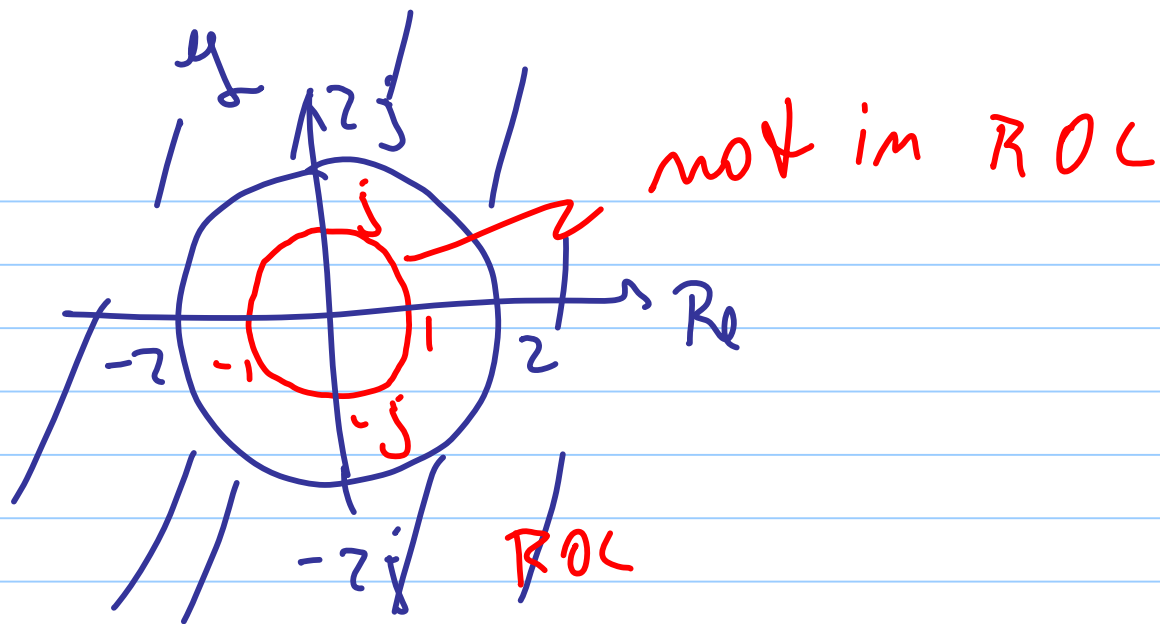
FACT: $X(\omega) = X(z) \Big|_{z = e^{j\omega}}$

provide $e^{j\omega} \in \text{ROC}$
"↑"
"belongs to"

Ex: $x[n] = 2^n u[n]$

$$X(z) = \frac{z}{z-2}, \quad \underbrace{\text{ROC } |z| > 2}$$

$$z = e^{j\omega} \rightarrow |z| = 1 \notin \text{ROC}$$



Meaning: for $x[n] = 2^n u[n]$

$$X(z) = \frac{z}{z-2} \quad |z| > 2$$

$X(\omega)$ does not exist

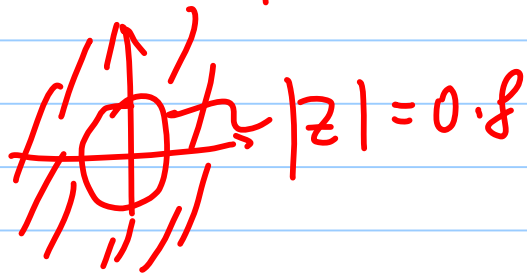
$$\{x\}: x[m] = 0.8^m u[m]$$

$$X(z) = \sum_{m=0}^{+\infty} 0.8^m z^{-m}$$

$$= \sum_{m=0}^{+\infty} \underbrace{(0.8z^{-1})^m}_{d} = \frac{z}{z - 0.8}$$

$$d = 0.8z^{-1}$$

ROC $|0.8z^{-1}| < 1 \Rightarrow |z| > 0.8$



$$X(\omega) = \sum_{m=0}^{\infty} 0.8^m e^{-j\omega m}$$

$$z \rightarrow e^{j\omega}, \quad |z| = 1$$

ROC: $|z| > 0.8$

unit circle ($|z| = 1$) \in ROC

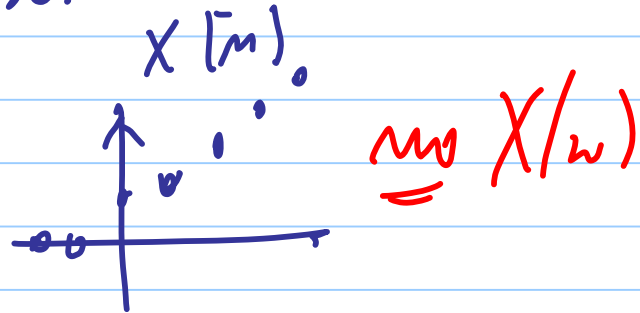
Since $X(z)$ converges in $|z| = 1$

$$X(\omega) = X(z) \Big|_{z=e^{j\omega}} \longrightarrow$$

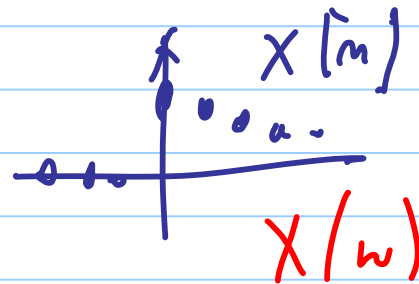
$$X(\omega) = \frac{e^{j\omega}}{e^{j\omega} - 0.8}$$

Compare the two cases:

① $x[n] = 2^n u[n]$



② $x[n] = 0.8^n u[n]$



① $n\omega X(\omega)$

② $X(\omega) = X(z) \Big|_{z=e^{j\omega}}$

FACT: condition for existence
of $X(w)$:

$$\sum_{m=-\infty}^{+\infty} |x[m]| < +\infty$$

convergent

Ex: ①

$$\sum_{m=0}^{+\infty} 2^m \rightarrow +\infty$$

not
conv.

②

$$\sum_{m=0}^{+\infty} 0.8^m = \frac{1}{1-0.8} = 5 < +\infty$$

conv.

Back to LDE:

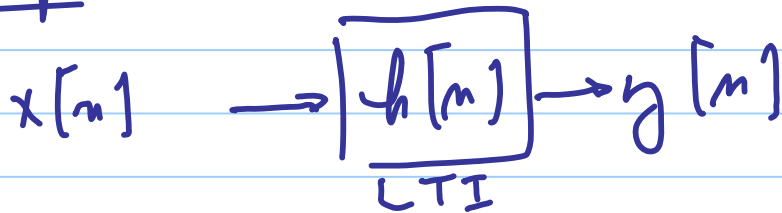
Properties of z-TRANSFORM

$$(1) \mathcal{Z}\{x[n-L]\} = z^{-L} X(z)$$

$L = \text{integer shift}$

$$(2) \mathcal{Z}\{h[n] * x[n]\} = H(z) X(z)$$

Proof: in the book.



LDE:

$$y[n] + a_1 y[n-1] + \dots + a_N y[n-N] =$$

$$= b_0 x[n] + \dots + b_N x[n-N]$$

apply z -Transform on both sides:

$$Y(z) + a_1 z^{-1} Y(z) + \dots + a_N z^{-N} Y(z) =$$

$$= b_0 X(z) + b_1 z^{-1} X(z) + \dots + b_N z^{-N} X(z)$$

some algebra

$$Y(z) = H(z)X(z)$$

$$\begin{aligned} H(z) &= \frac{b_0 + b_1 z^{-1} + \dots + b_N z^{-N}}{1 + a_1 z^{-1} + \dots + a_N z^{-N}} = \\ &= \frac{b_0 z^N + b_1 z^{N-1} + \dots + b_N}{z^N + a_1 z^{N-1} + \dots + a_N} \end{aligned}$$

① $H(z)$ = transfer function

② $H(z) = \mathcal{Z}\{h[n]\}$, $h[n]$ impulse
response

Example :

$$y[n] + 0.8y[n-1] = 2x[n-1] \quad \checkmark$$

$$H(z) = \frac{2z^{-1}}{1 + 0.8z^{-1}} =$$

$$= \frac{2}{z + 0.8}$$

TRANSFER FUNCTION

FREQUENCY RESPONSE?

Recall: $H(\omega) = \text{DTFT}\{h[n]\}$
definition

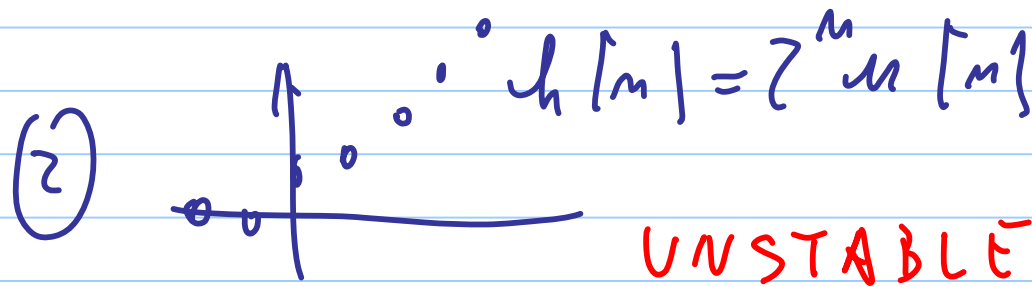
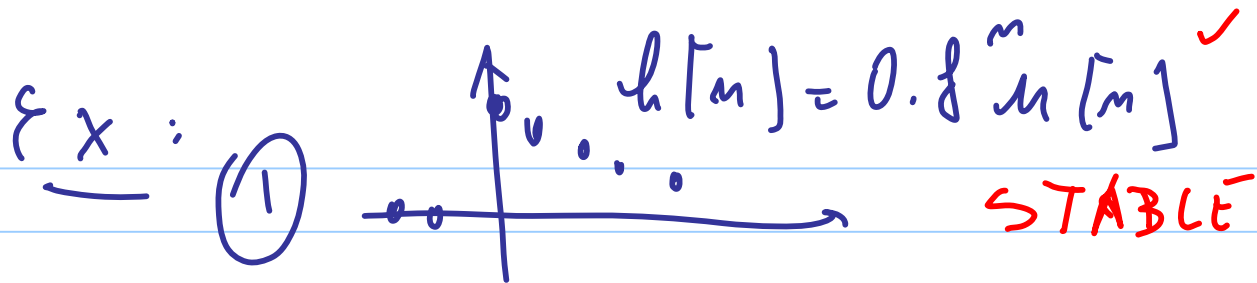
$$\begin{aligned} x[n] &= e^{j\omega n} \\ &\rightarrow \boxed{H(\omega)} \rightarrow y[n] = \underbrace{H(\omega)}_{\substack{\text{effect of} \\ \text{system on} \\ \text{mag. and phase}}} e^{j\omega n} \end{aligned}$$

Freq. Response from Transfer function
what we have seen :

$$H(\omega) = H(z) \Big|_{z = e^{j\omega}}$$

provided $\sum_{n=-\infty}^{+\infty} |h[n]| < \infty$

if not $H(\omega)$ does not exist

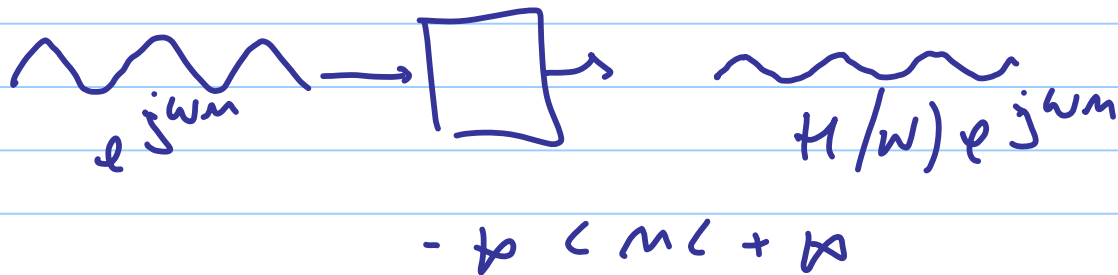


FACT: if system stable
 $H(\omega) = H(z) \big|_{z=e^{j\omega}}$;
 if system unstable
 no $H(\omega)$ freq. response.

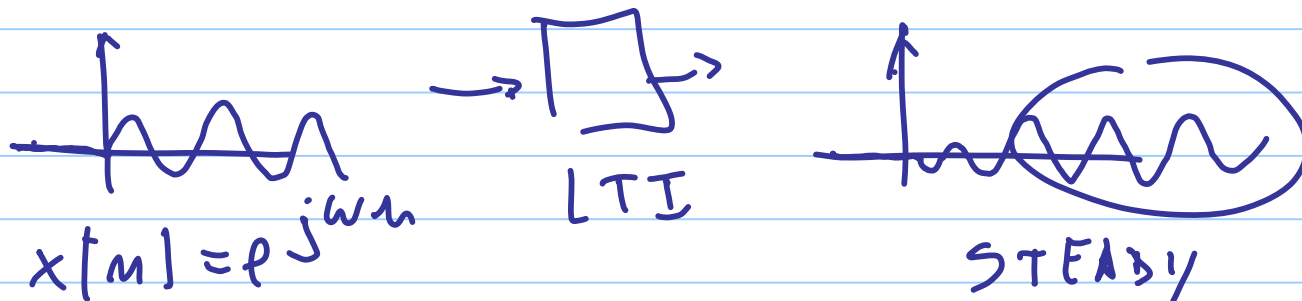
DFT vs Z-Transform (FT vs. Laplace)

Freq. response: steady state

Theory:



In practice: if system stable

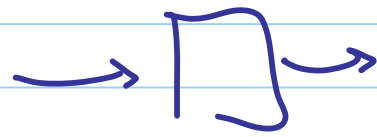
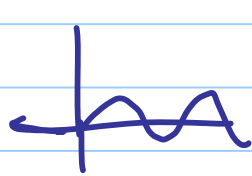


$$y[n] = H(\omega) e^{j\omega n}$$

STEADY STATE EXISTS \Rightarrow

\Rightarrow FREQ. RESP. EXISTS

If system is unstable



no steady state

no Freq. Resp.

no $H(\omega)$

FACT: stability of causal systems

$$H(z) = \frac{B(z)}{A(z)}, \quad B(z), A(z) \text{ polynomials}$$

FACTOR $B(z), A(z)$ into roots:

$$H(z) = K \frac{(z-z_1) \dots (z-z_M)}{(z-p_1) \dots (z-p_N)}$$

$z_i = \text{ZEROS}$

$p_i = \text{poles.}$

Ex:

$$y[n] + y[n-1] + y[n-2] = 2x[n-1] - x[n-2]$$

$$H(z) = \frac{2z^{-1} - z^{-2}}{1 + z^{-1} + z^{-2}} =$$

$$= \frac{2z - 1}{z^2 + z + 1} = \frac{B(z)}{A(z)}$$

$$z_i \Rightarrow 2z - 1 = 0 \Rightarrow z_i = \frac{1}{2} \quad \underline{\underline{z = 0.5}}$$

$$p_i \rightarrow z^2 + z + 1 = 0 \Rightarrow p_{1,2} = \frac{-1 \pm \sqrt{1-4}}{2} =$$
$$= \left(-0.5\right) \pm j \left(\frac{\sqrt{3}}{2}\right) \quad \underline{\underline{\text{poles}}}$$

$$= -0.5 \pm j 0.866$$

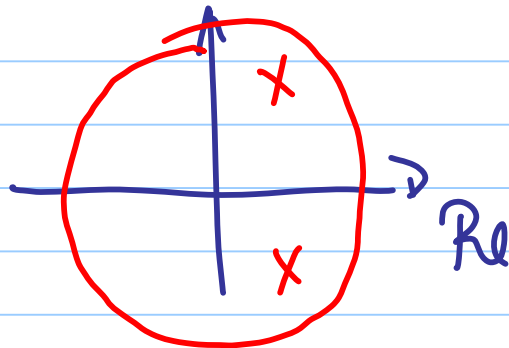
Then

$$H(z) = 2 \frac{z - 0.5}{(z + 0.5 - j0.866)(z + 0.5 + j0.866)}$$

FACT: a causal system is stable
if and only if

$$|p_i| < 1$$

, i.e. inside
unit circle.



For example:

$$|p_i| = \sqrt{0.5^2 + \frac{3}{4}} = 1$$

not inside the unit circle
i.e. system unstable

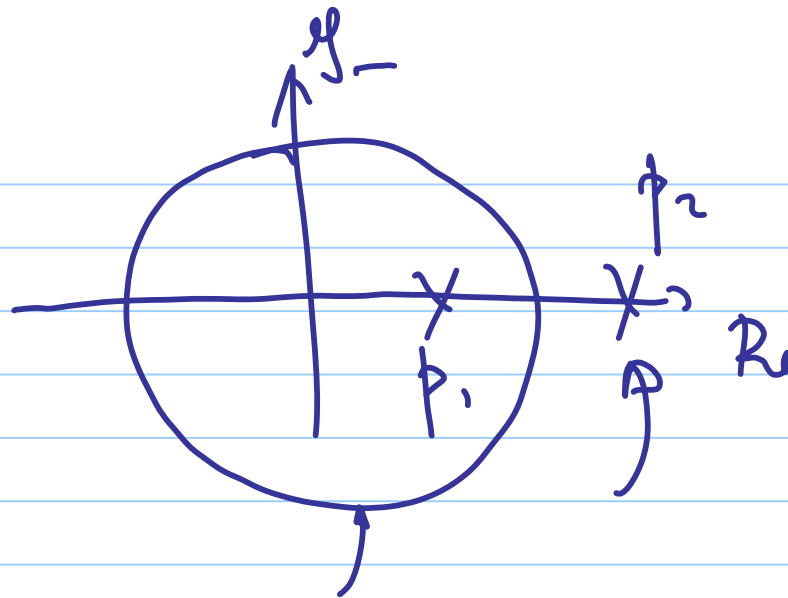
Ex:

$$y[n] + 2y[n-1] + 0.8y[n-2] = x[n-1]$$

$$H(z) = \frac{z^{-1}}{1 + 2z^{-1} + 0.8z^{-2}} =$$

$$= \frac{z}{z^2 + 2z + 0.8}$$

$$p_i: z^2 + 2z + 0.8 = 0, \quad p_{i,2} = \frac{-2 \pm \sqrt{4 - 0.64}}{2}$$
$$= -1 \pm (\dots)$$

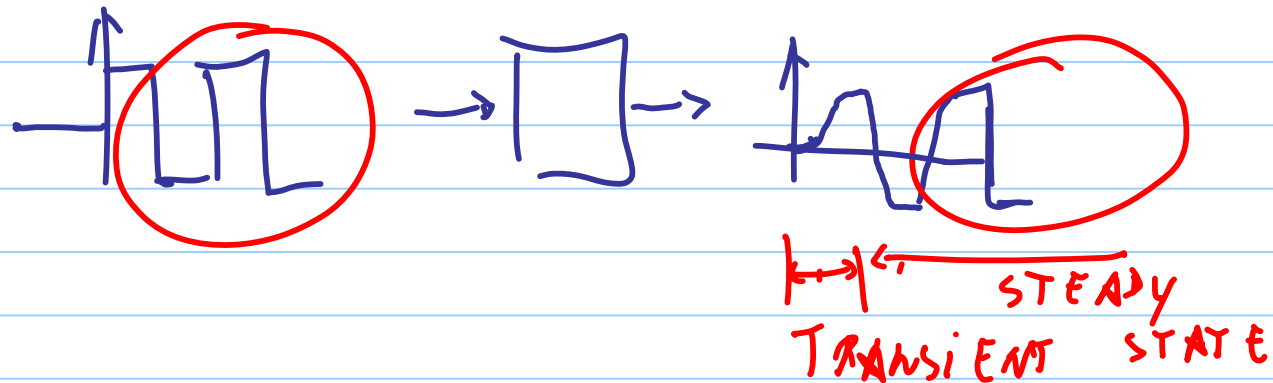


unit circle

Since at least one pole outside
unit circle \Rightarrow system unstable

For stable systems: z-Transform vs DTFT
(LT vs FT)

① steady state: DTFT



② Transfer function $H(z)$
transient & steady state

